

# Bidirectional Double Buck Boost Dc- Dc Converter

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**Abstract**— For renewable energy resources a new double buck boost coupled inductor based bidirectional converter is presented. With simple circuit high conversion ratio achieved. In order to achieve a high step up conversion ratio by controlling one power switch, during discharging mode, its acts as a double boost converter. Similarly in order to achieve a high step down conversion ratio by controlling two powers switches simultaneously during charging mode, its acts as a double buck converter. This two happen in high ratio as double buck boost is enabled. The advantage of this converter is energy stored in coupled inductor is recycled in order reduce leakage inductance; reduce switching loss, high voltage stress, hence we can achieve high efficiency. MATLAB is used to simulate the circuit and hardware prototype will do in open loop.

**Key words:** Bidirectional Converter, Coupled Inductor, High Conversion Ratio

## I. INTRODUCTION

The bidirectional dc-dc converter has wide range of applications in battery charging, UPS, and hybrid vehicles. UPS is used for computers, telecommunication equipments, and electronic instruments. It is able to transfer or balanced energy between load and battery. When availability of renewable energy, energy is transferring from sources to the load and also transfers part of energy to the battery. During utility failure or renewable energy is insufficient condition, the battery is transferred its stored energy continuously without interrupt to the load. The battery is used as a storage device or back up whenever utility failures. Fig.1.1.shows the block diagram Renewable energy hybrid system

The conventional buck-boost dc-dc converter is not able to provide high step up/down voltage gain ratio due to the presence of parasitic elements. But it has advantage like simple configuration. The other converters like forward fly back converters/ fly back converter, half bridge, full bridge converter, multilevel, switched capacitor type etc are have some disadvantages, like we have to adjust the turn ratio of the transformer and proper duty cycle in order to get high voltage gain ratio. Conventional buck-boost dc-dc converters are used only in low power applications. Leakage inductance, high current stress and conduction loss, high voltage spike on the power switch due to stored energy in inductor. Control circuit is also complicated.

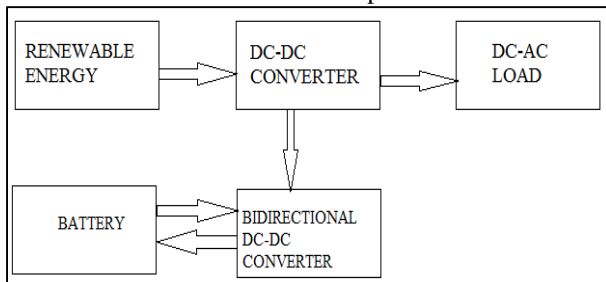


Fig.1.1.Renewable energy hybrid system

In order to overcome these problems, the proposed bidirectional double buck boost dc-dc converter is used to achieve high voltage gain ratio. The coupled inductor technique which is used to reduces the leakage inductance current stress and conduction losses by recycled leakage inductance energy of the coupled inductor. It is also improves the efficiency by providing low  $R_{DS-ON}$  switch.

## II. OPERATING PRINCIPLES OF THE PROPOSED CONVERTER

The proposed converter is able to transfer energy between two different dc sources. Such has low voltage side voltage 24V to the high voltage side 200V and the output power of 200W. It has two different modes such as discharging and charging mode is explain by its equivalent circuit diagram. Fig.2.1shows the proposed converter circuit with leakage inductances.

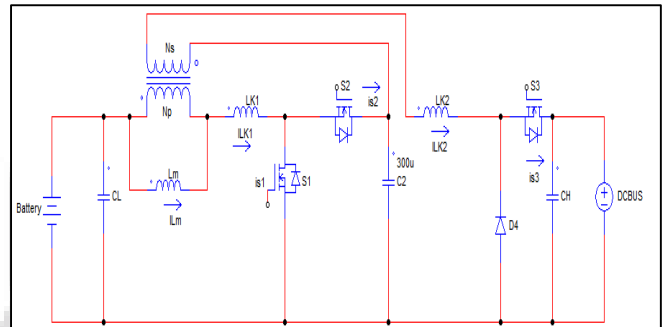


Fig. 2.1:Proposed converter circuit with leakage inductances

### A. Discharging mode

In discharging mode converter acts as a two stage double boost converter by controlling the power switch  $S_1$ . The switch  $S_1$  is the main power switch. The switches  $S_2$  and  $S_3$  are not conduct during the entire period. The explanations of different modes are described as follows.

**Mode 1:** During this mode 1, switch  $S_1$  and diode  $D_{S3}$  starts conducting. It is shows in below Fig.2.2 by its equivalent circuit. The Leakage inductor  $L_{K2}$  stored some energy that will transfer to  $C_H$  via  $i_{D3}$ , hence stored energy is gradually reduce in  $C_2$ ,  $L_{K2}$ ,  $I_{S3}$ .The battery transfer its energy into leakage inductor  $L_{K1}$ , hence current and its energy in the leakage inductor is slowly increases. The mode will be end when current  $I_{S3}$  comes to zero and diode  $D_{S3}$  is stopps conducting.

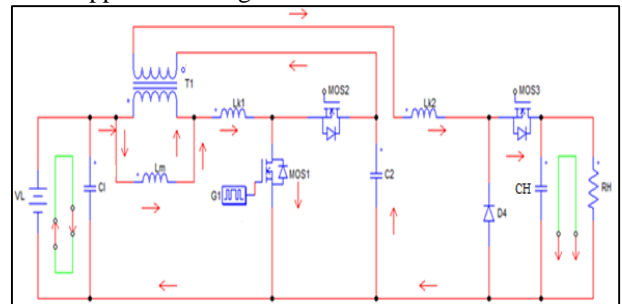


Fig.2.2 Equivalent circuit mode 1

**Mode 2:** During this mode switch  $S_1$  and diode  $D_4$  are starts conducting. It is show in below Fig.2.3.by its equivalent circuit. Battery is continuously charging the magnetizing inductor  $L_m$  and the leakage inductor  $L_{K1}$  during this mode. Hence current flowing through the magnetizing-inductor current  $i_{Lm}$  and the leakage-inductor current  $i_{L_{k1}}$  are linearly increased.  $C_2$  get energy from  $V_L$  through  $N_S$  and  $D_4$ .the voltage appears across  $C_2$  is equal to  $nV_L$ . This mode will end when  $s_1$  is stops conducting.

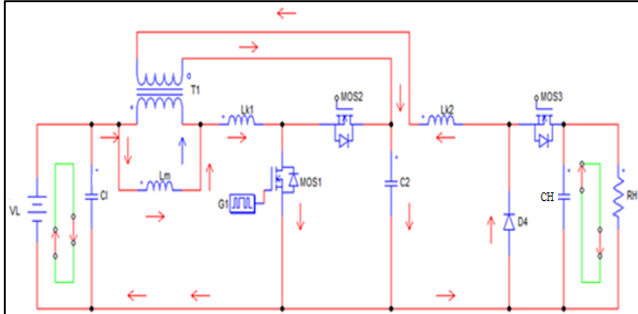


Fig .2.3.Equivalent circuit mode 2

**Mode 3:** During mode 3,  $S_1$  and  $D_{S3}$  are stops conducting and  $D_{S2}$  starting conducting. It is shown in below Fig.2.4.by its equivalent circuit diagram. The  $L_{K1}$  and  $L_{K2}$  transfer its energy into  $C_2$  through  $D_{S2}$  and  $D_4$ , respectively. The mode will be end when current  $I_{LK2}$  flowing through  $I_{d4}$  is equal to zero and  $D_4$  stops conducting.

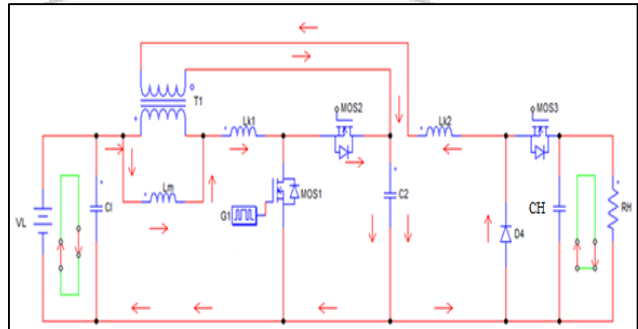


Fig. 2.4.Equivalent circuit mode 3

**Mode 4:** During mode 4, when  $S_1$  is stops conducting, and  $D_{S2}$  and  $D_{S3}$  are starts conducting. It is show in below Fig.2.5.by its equivalent circuit diagram. The capacitor  $C_2$  is get energy by  $V_L, L_m$  and  $L_{K1}$  through  $D_{S2}$ .the  $L_m$  transfer its energy to load capacitor  $C_H$  and load resistance  $R_H$  through  $L_{K2}$ . The mode will be end when voltage at,  $C_2 = nV_{in}$ .

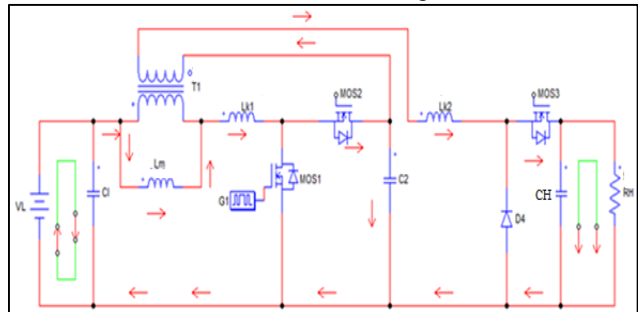


Fig. 2.5.Equivalent circuit mode 4

**Mode 5:** During mode 5,  $S_1$  is stops conducting, and  $D_{S2}$  and  $D_{S3}$  are starts conducting. It is shown in below Fig.2.6.by its equivalent circuit diagram. The energy transferring into  $C_H$  and  $R_H$  by  $L_m$  through  $L_{K1}$ ,  $L_{K2}$  and also  $D_{S3}$ .this mode end when  $i_{L_{k1}}$  is equal to zero.

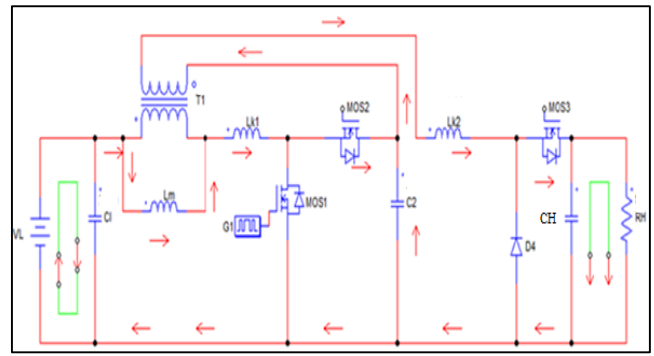


Fig. 2.6.Equivalent circuit mode 5

**Mode 6:** During mode 6,  $S_1$  and  $D_{S2}$  are stopping conducting, and  $D_{S3}$  is starts conducting. It is show in below Fig.7.by its equivalent circuit diagram. The energy stored in  $L_m$  transfer into  $C_H$  and  $R_H$  through  $L_{K2}$  and  $D_{S3}$ . The energy stored in  $C_2$  is also transferred to  $C_H$  and  $R_H$ .the mode will be end when  $S_1$  starts conducting.

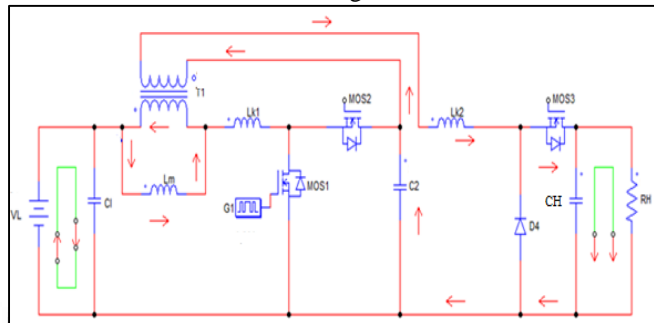


Fig. 2.7.Equivalent circuit mode 6

The voltage gain of the discharging mode is given as  $VH/VL = n/1 - D$ . (2.1)

**B. Charging mode:**

In charging mode power switches  $s_1$  and  $s_2$  is conduct simultaneously and  $s_1$  is switch off for all modes. The explanations of different modes are described as follows.

**Mode 1:** During mode 1, the diode  $D_{S1}$  is starts conducting. It is shows in below fig.2.8.by its equivalent circuit diagram. Load capacitor  $C_L$  and load resistance  $R_L$  is get energy from  $L_m$ .Hence current in  $L_m$  in slowly decreases. The  $L_{K2}$  transfer energy to capacitor to  $C_2$  which is recycled through  $D_4$ .The mode will be end when current  $I_{D4}$  comes to zero. Thus,

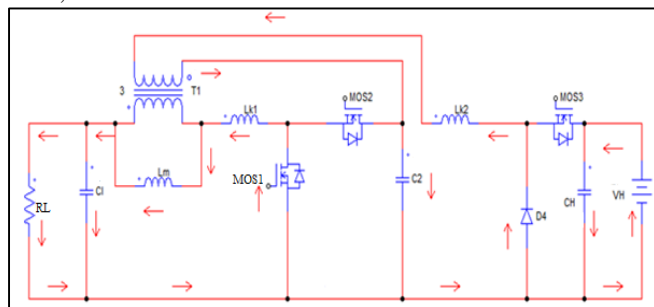


Fig. 2.8.Equivalent circuit mode 1

**Mode 2 :**During mode 2,the switches  $S_2$  and  $S_3$  are starts conducting,it is shows in below fig.2.9.by its equivalent circuit.The  $V_H$  transfer its energy to  $C_2, C_L, R_L$  and also transfer energy to  $L_m$ ,  $L_m$  start charging,the mode will be end when capacitor  $C_2$  transfer its energy to  $R_L$ .

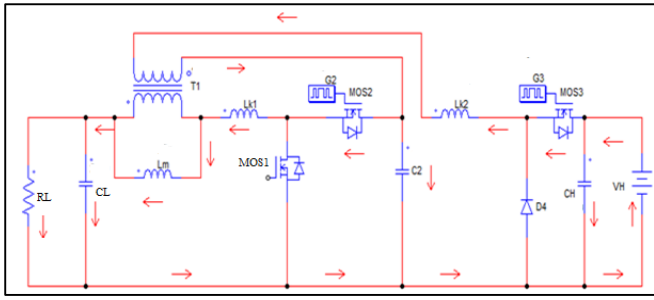


Fig.2.9.Equivalent circuit mode 2

**Mode 3:** During mode 3, the switches  $S_2$  and  $S_3$  still conducting. It is shows in below fig.2.10.by its equivalent circuit diagram.  $V_H$  and  $C_2$  transfer its energy to  $L_m$ ,  $C_L$  and  $R_L$ .  $L_m$  gets energized slowly by transferred energy. The mode will be ends when  $S_2$  and  $S_3$  are switch off.

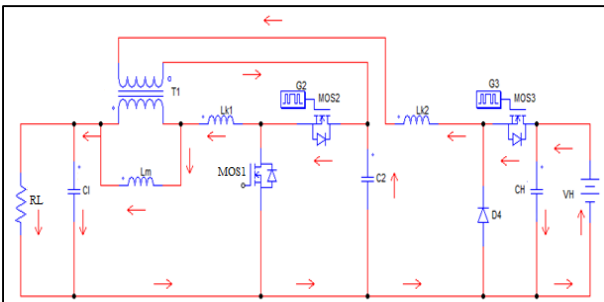


Fig.2.10.Equivalent circuit mode 3

**Mode 4:** During mode 4, switches  $S_2$  and  $S_3$  are switch off and  $D_{S1}$  is switch on .It is shows in below fig.2.11.by its equivalent circuit diagram. The energy stored in inductor  $L_{K1}$ , transfer its energy to  $C_L$  and similarly  $L_{K2}$  transfer energy to  $C_2$ .the mode will be end, when stored energy in  $L_{K2}$  completely transfer to zero. Thus,

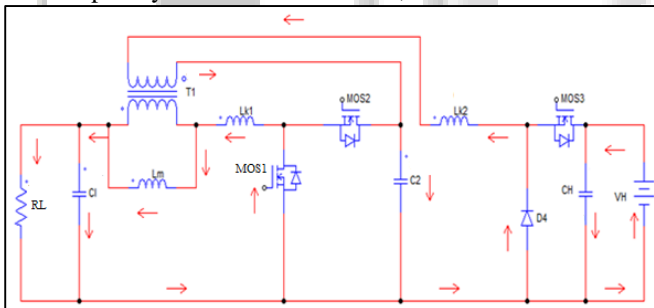


Fig.2.11. Equivalent circuit mode 4

**Mode 5:** During mode 5, the switches  $S_2$  and  $S_3$  are switch off and  $I_{D_{S1}}$  is still conducting. It is shows in below fig.2.12.by its equivalent circuit diagram. The  $L_m$  transfer its energy to both  $C_L$  and  $R_L$ .  $C_2$  get energy via  $N_S$  and  $D_4$ .

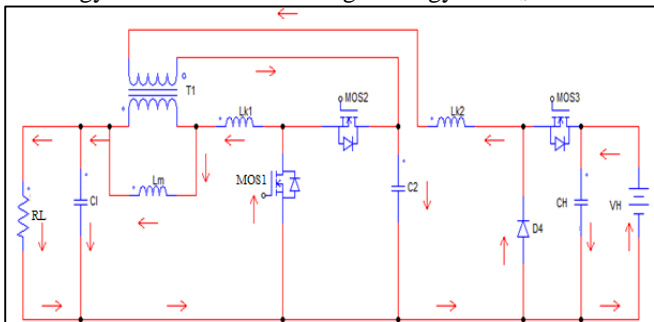


Fig.2.12. Equivalent circuit mode 5

The voltage gain of the charging mode is given as  $V_L/V_H = D/1 + n - nD$ . (2.2)

### III. SIMULATION RESULTS

The operation of different modes in discharging and charging mode is learned by above its equivalent circuit model. We have get simulation results by Simulink model and converter parameters are give in below table 1.

Symbols	Names	Values
$V_L$	Input DC voltage	24V
$V_H$	Output DC voltage	200V
A	voltage gain	8.33
$f_{sw}$	Switching frequency	50k
$L_m$	Magnetizing inductance	37mh
$N_P$ and $N_S$	Turns ratio	1:3
L	Inductor ( $L_{K1}$ & $L_{K2}$ )	0.33uH
C	Capacitor ( $C_2$ & $C_H$ )	300uF
$C_L$	Load capacitance	220uF
D	Discharging mode	67%
D	Charging mode	35.5%
$P_{out}$	Output power	200W

Table .1: Parameters for Simulation

### IV. DISCHARGING MODE

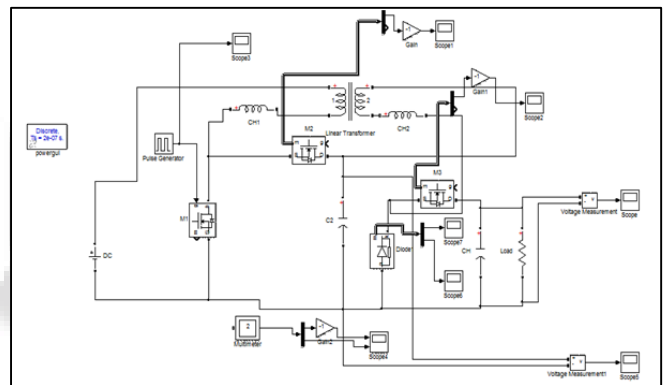


Fig. 3.1. Simulation circuit for discharging mode

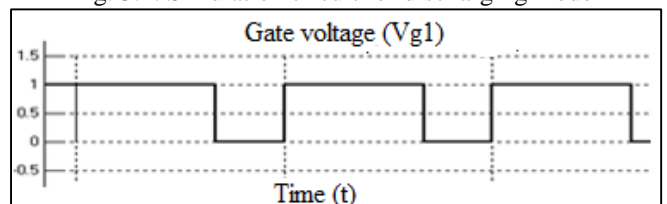


Fig. 3.2.Gate voltage  $V_{G1}$

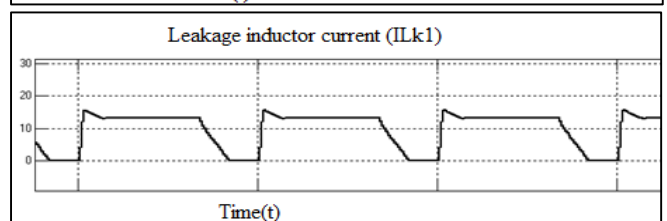
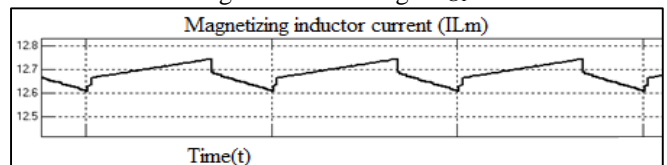


Fig. 3.3.Magnetizing current ( $I_{Lm}$ ) and leakage current ( $I_{Lki}$ )

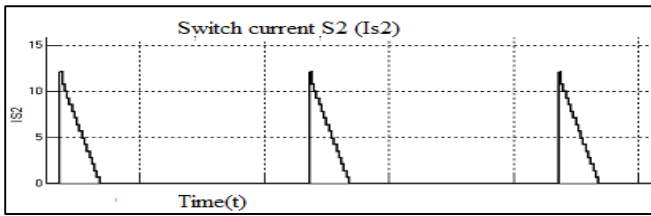


Fig. 3.4. Switch current  $S_2$  ( $I_{S2}$ )

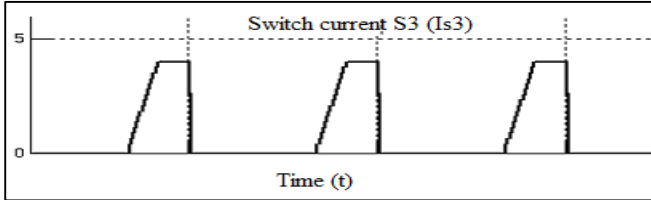


Fig. 3.5. Switch current  $S_3$  ( $I_{S3}$ )

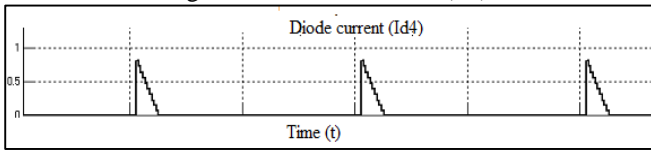


Fig. 3.6. Diode current ( $I_{D4}$ )

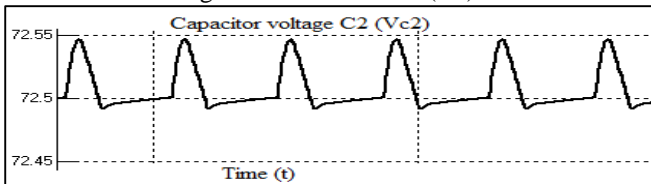


Fig. 3.7. Capacitor voltage  $C_2$  ( $V_{C2}$ )

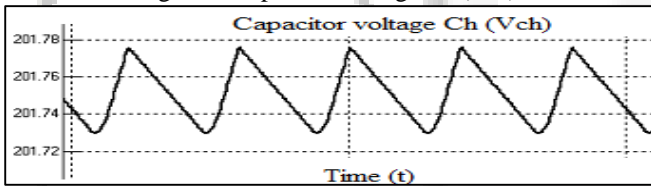


Fig. 3.8. Capacitor voltage  $C_H$  ( $V_{CH}$ )

A. Charging mode

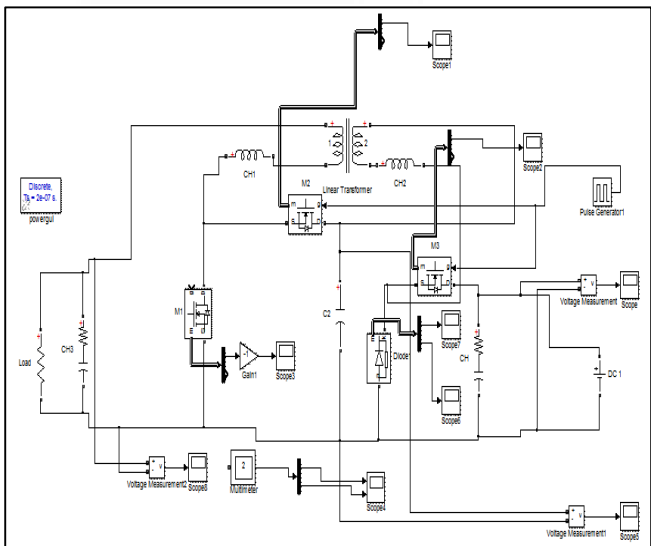


Fig. 3.9. Simulation circuit for charging mode

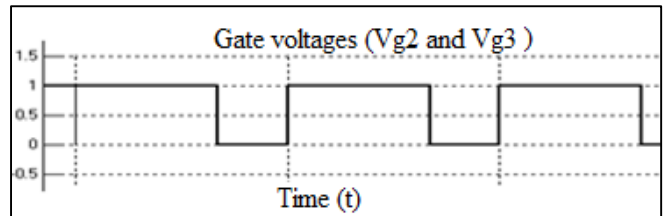


Fig. 3.10. Gate voltages  $V_{G2}$  and  $V_{G3}$

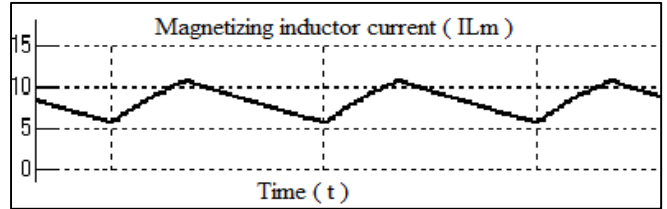


Fig. 3.11. Magnetizing inductor current ( $I_{Lm}$ )

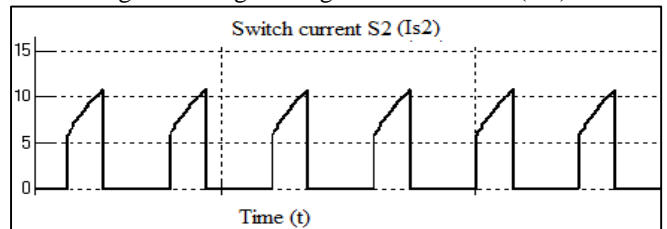


Fig. 3.12. Switch current ( $I_{S2}$ )

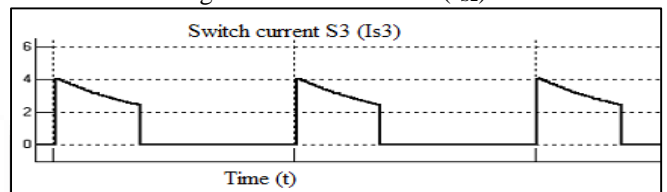


Fig. 3.13. Switch current ( $I_{S3}$ )

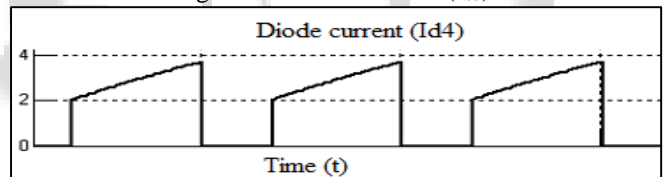


Fig. 3.14. Diode current ( $I_{D4}$ )

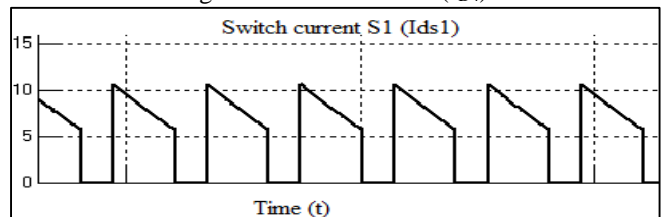


Fig. 3.15. Switch current ( $I_{Ds1}$ )

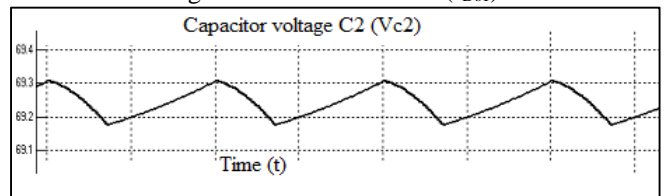


Fig. 3.16. Capacitor voltage ( $V_{C2}$ )

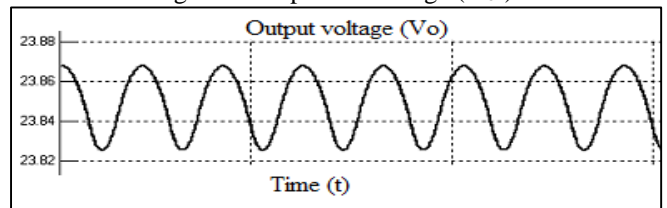


Fig. 3.17. Output voltage ( $V_o$ )



B. Closed loop operation for discharging mode:

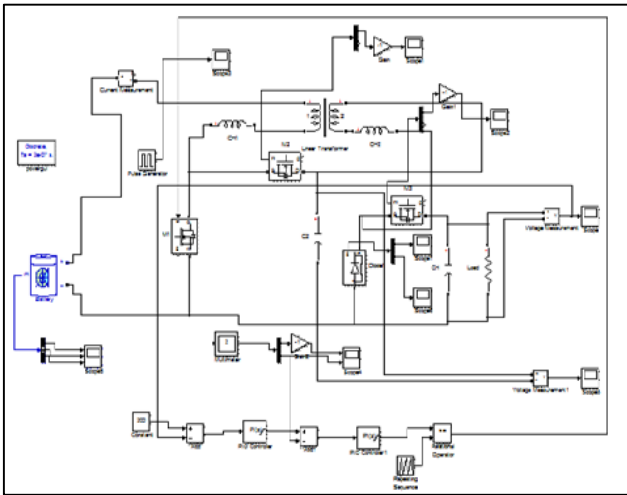


Fig. 3.18. Close loop simulation module for discharging mode

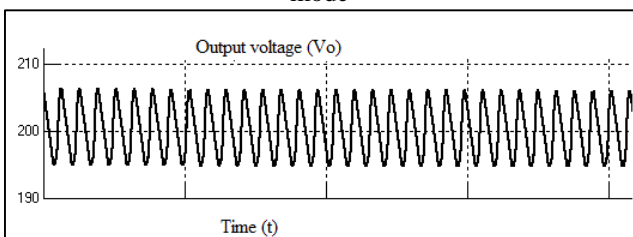


Fig.3.19. Output voltage in close loop

C. Closed loop operation for charging mode:

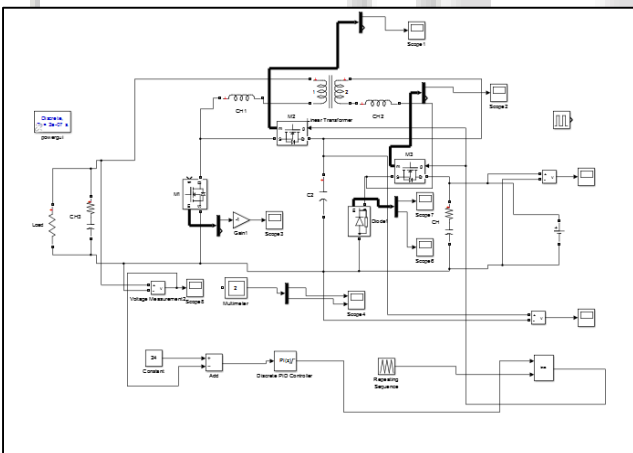


Fig. 3.20. Close loop simulation module for charging mode

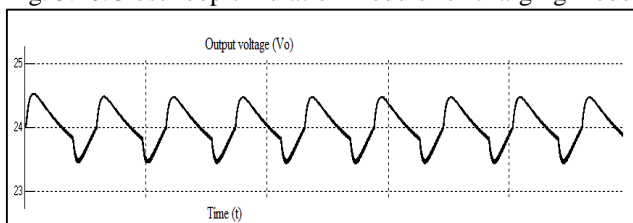


Fig. 3.21. output voltage for charging mode

V. CONCLUSION

This paper presents a bidirectional double buck boost dc -dc converter for renewable energy systems. The coupled inductor technique is used to achieve high step up/down conversion ratio during discharging and charging operation. The leakage inductance, high voltage spike in power switches, high current stress and conduction loss can be

reduce by using couple inductor technique. This converter can be used for high voltage applications. High efficiency is achieved by recycled the coupled inductor energy.

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