

Design and Implementation of Axi-APB Bridge based on Amba 4.0

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Abstract— ARM introduced the Advanced Microcontroller Bus Architecture (AMBA) 4.0 specifications in March 2010, which includes Advanced eXtensible Interface (AXI) 4.0. AMBA bus protocol has become the de facto standard SoC bus. That means more and more existing IPs must be able to communicate with AMBA 4.0 bus. Based on AMBA 4.0 bus, we designed an Intellectual Property (IP) core of Advanced Peripheral Bus (APB) bridge, which translates the AXI4.0-lite transactions into APB 4.0 transactions. The bridge provides an interface between the high-performance AXI bus and low-power APB domain.

Key words: SoC, AMBA, AXI, APB

I. INTRODUCTION

Integrated circuits has entered the era of System-on-a-Chip (SoC), which refers to integrating all components of a computer or other electronic system into a single chip. It may contain digital, analog, mixed-signal, and often radio-frequency functions – all on a single chip substrate. With the increasing design size, IP is an inevitable choice for SoC design. And the widespread use of all kinds of IPs has changed the nature of the design flow, making On-Chip Buses (OCB) essential to the design. Of all OCBs existing in the market, the AMBA bus system is widely used as the de facto standard SoC bus. On March 8, 2010, ARM announced availability of the AMBA 4.0 specifications. As the de facto standard SoC bus, AMBA bus is widely used in the high-performance SoC designs. The AMBA specification defines an on-chip communication standard for designing high-performance embedded microcontrollers. The AMBA 4.0 specification defines five buses/interfaces.

- Advanced eXtensible Interface (AXI)
- Advanced High-performance Bus (AHB)
- Advanced System Bus (ASB)
- Advanced Peripheral Bus (APB)
- Advanced Trace Bus (ATB)

AXI, the next generation of AMBA interface defined in the AMBA 4.0 specification, is targeted at high performance, high clock frequency system designs and includes features which make it very suitable for high speed sub-micrometer interconnect.

- separate address/control and data phases
- Support for unaligned data transfers using byte strobes
- burst based transactions with only start address issued
- issuing of multiple outstanding addresses
- easy addition of register stages to provide timing closure

II. TOP VIEW

A. Block Diagram

In this study, we focused mainly on the implementation aspect of an AXI4-Lite to APB bridge. The APB bridge provides an interface between the high-performance AXI domain and the low-power APB domain. It appears as a slave on AXI bus but as a master on APB that can access up to sixteen slave peripherals. Read and write transfers on the AXI bus are converted into corresponding transfers on the APB. The AXI4-Lite to APB bridge clock diagram is shown in Figure. 1.

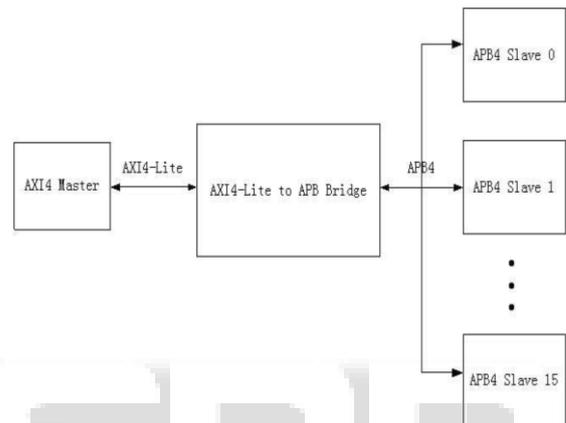


Fig. 1: block diagram

B. Signal Connections

Figure. 2 shows the component signal connections. The bridge uses:

- AMBA AXI-Lite signals as described in the AMBA AXI-Lite 4.0 protocol specification.
- AMBA APB signals as described in the AMBA APB 4.0 protocol specification.



C. AXI Handshake Mechanism

In AXI 4.0 specification, each channel has a VALID and READY signals for handshaking. The source asserts VALID when the control information or data is available. The destination asserts READY when it can accept the control information or data. Transfer occurs only when both the VALID and READY are asserted. Figure. 3 shows all possible cases of VALID/READY handshaking. Note that when source asserts VALID, the corresponding control information or data must also be available at the same time. The arrows in Figure. 3 indicate when the transfer occurs. A transfer takes place at the positive edge of clock. Therefore, the source needs a register input to sample the READY signal. In the same way, the destination needs a register input to sample the VALID signal. Considering the situation of Figure. 3(c), we assume the source and destination use output registers instead of combination circuit, they need one cycle to pull low VALID/READY and sample the VALID/READY again at T4 cycle. When they sample the VALID/READY again at T4, there should be another transfer which is an error. Therefore source and destination should use combinational circuit as output. In short, AXI protocol is suitable register input and combinational output circuit.

The APB bridge buffers address, control and data from AXI4-Lite, drives the APB peripherals and returns data and response signal to the AXI4-Lite. It decodes the address using an internal address map to select the peripheral. The bridge is designed to operate when the APB and AXI4-Lite have independent clock frequency and phase. For every AXI channel, invalid commands are not forwarded and an error response generated. That is once a peripheral accessed does not exist, the APB bridge will generate DE CERR as response through the response channel (read or write). And if the target peripheral exists, but asserts PSLVERR, it will give a SLVERR response.

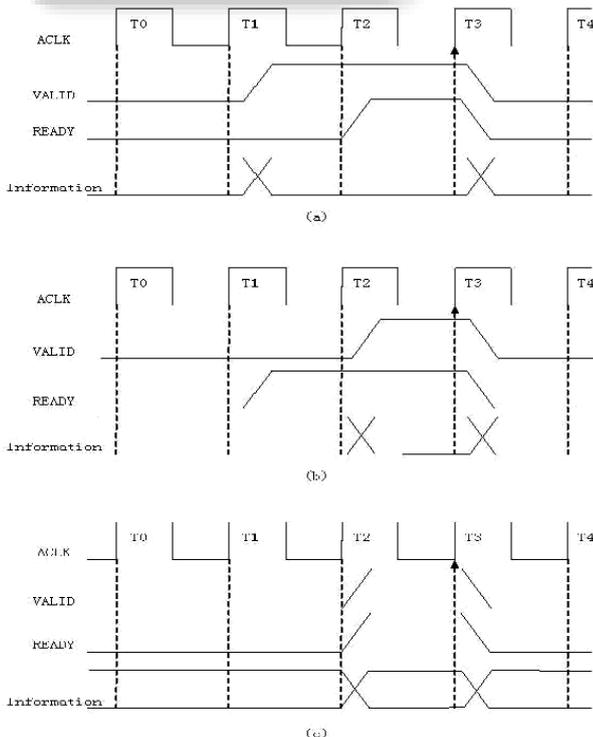


Fig. 3: handshake mechanism

III. CLOCK DOMAIN AIN CROSSING (CDC)

A clock domain crossing (CDC), or simply clock crossing, is when a signal crosses from one clock domain into another^[3]. If a signal does not assert long enough and is not registered, it may appear asynchronous on the incoming clock boundary. Metastability happens when signal changes within the setup/hold time window. Synchronizing a signal that crosses into a higher clocked domain can be accomplished by registering the signal through a flip-flop that is clocked by the source domain, thus holding the signal long enough to be detected by the higher clocked destination domain. Synchronizing a signal traversing into a slower clock domain is more cumbersome^[4]. This typically requires a register in each clock domain with a form of feedback from the destination domain to the source domain, indicating that the signal was detected.

A. Metastability

Metastability cannot be avoided, but a solution for handling the metastable signal enables proper functioning of the design. The metastability occurrences can be predicted by using the mean time between failures (MTBF) formula (1): Where C1 and C2 are constants that depend on the technology used to build the flip-flop; tMET is the duration of the metastable output; and fclk and fdata are the frequencies of the synchronous clock and the asynchronous input, respectively.

B. Synchronizer

Designers can use special metastable hardened flops for increasing the MTBF. For example, in Figure. 4, a synchronizer flop is used following the signal DB. So, instead of the metastable signal DB being used in the function downstream as in Figure. 4, the stable signal DB2 is used in the logic downstream^[8]. In the AXI4-Lite to APB bridge, we use synchronizer block designs for communicate between the AXI and APB clock domain.

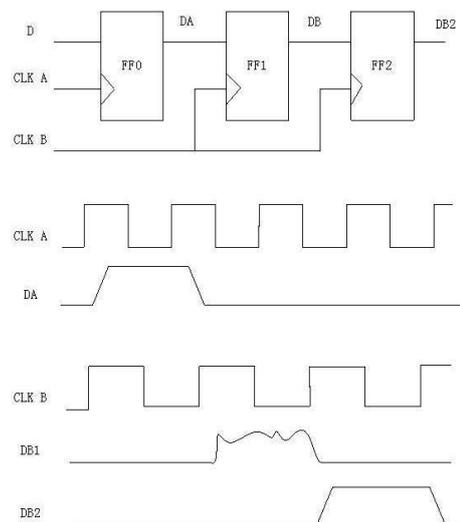


Fig. 4: two flip-flop synchronizer

IV. FINITE STATE MACHINE (FSM)

A finite state machine is a mathematical abstraction sometimes used to design digital logic or computer programs. It is a behavior model composed of a finite

number of states, transitions between those states, and actions, similar to a flow graph in which one can inspect the way logic runs when certain conditions are met^[10]. The state transition diagram is a picture of our state machine model. Figure. 5 is the state transition diagram of our FSM.

The state machine operates through the following states:

- IDLE. This is the default state of the FSM.
- WRITE_SETUP. When a write transfer request is asserted, the FSM moves into the WRITE_SETUP state.
- READ_SETUP. When a read transfer request is asserted, the FSM moves into the READ_SETUP state.
- WRITE_ACCESS. The enable signal, PENABLE, is asserted in the WRITE_ACCESS state.
- READ_ACCESS. The enable signal, PENABLE, is asserted in the READ_ACCESS state.
- WRITE_WAIT. When the AXI write response channel is not ready for receiving signal BRESP, then stay in WRITE_WAIT state.
- READ_WAIT. When the AXI read data channel is not ready for receiving signal RRESP, then stay in READ_WAIT state.

States READ_WAIT and WRITE_WAIT are added, because the APB is not pipelined, wait states are added during transfers between the APB and AXI interface.

According AXI specification, the read address channel, write address channel and write data channel are completely independent. Each channel has a set of forward signals and a feedback signal for handshaking. A read and a write requests may be issued simultaneously (AWVALID/WVALID and ARVALID are asserted high simultaneously) from AXI4-Lite, the AXI4-Lite to APB bridge will give more priority to the read request than to the write request. That is, when both write and read requests are valid, the write request is initiated on APB after the read is requested on APB.

V. SIMULATION & IMPLEMENTATION

The timing diagram shown in Figure. 5 illustrate the AXI4-Lite to APB bridge operation for various read and write transfers. It shows that when both read and write requests are active, read is given more priority.

Before Static Timing Analysis (STA), it is necessary to inform the EDA tools that ACLK and PCLK are two asynchronous clock domains.

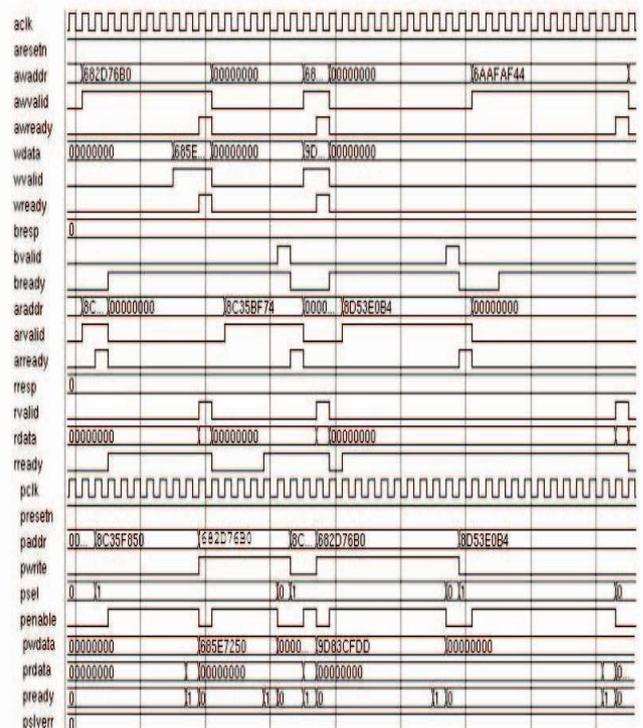


Fig. 5: typical read and write transfer

VI. CONCLUSION

In this study, we provide an implementation of AXI4-Lite to APB bridge which has the following features:

- 32-bit AXI slave and APB master interfaces.
- PCLK clock domain completely independent of ACLK clock domain.
- support up to 16 APB peripherals.
- support the PREADY signal which translate to wait states on AXI.
- an error on any transfer results in SLVERR as the AXI read/write response.

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